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Contact Author

Institut de RadioAstronomie Millimétrique

PdB Axis Velocity Loop Analysis

(and much more)

Owner Alain Perrigouard (perrigou@iram.fr)

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Approved by:

A.Perrigouard

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Content

- 1 Velocity servo loop step response..... 3**
- 1.1 Azimuth step response3
- 1.2 Elevation step response4
- 2 Velocity servo loop step response..... 5**
- 2.1 Azimuth step response5
- 2.2 Elevation step response6
- 3 Frequency analysis 7**
- 3.1 Azimuth:.....7
- 3.2 Bode Analysis. Data preprocessing:9
- 3.3 Transfer function.....10
- 4 Ideal case analysis 12**
- 5 Position servo loops 13**
- 6 System time delays23**

1 Velocity servo loop step response

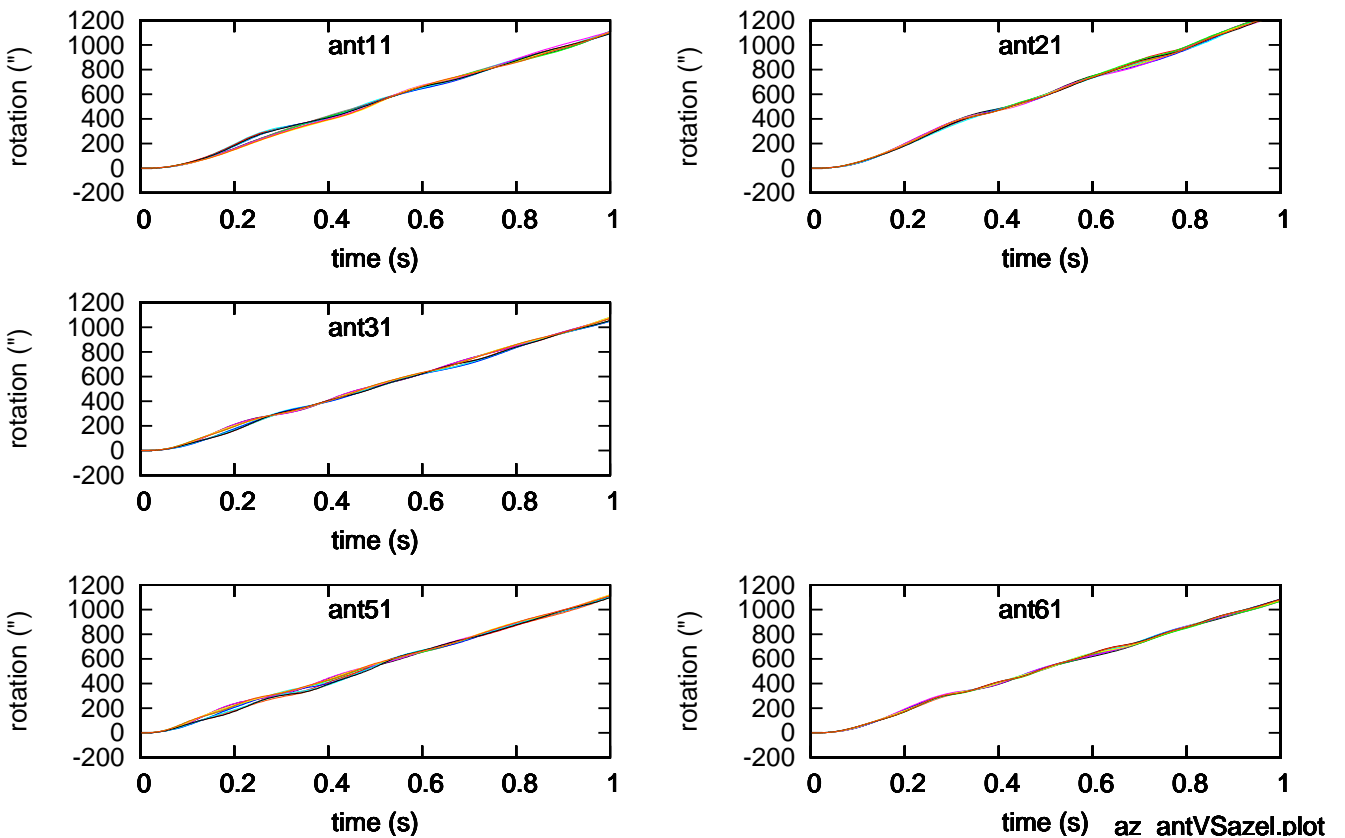
Versus several azimuth and elevation starting positions

Positive and negative steps of $\pm 3V$ have been applied to the azimuth and to the elevation servo velocity loops for different positions of the antennas. Az=0deg, 90deg, and -90deg. El=20deg, 45deg and 70deg.

1.1 Azimuth step response

The input DAC step command is equal to ± 10000 . The DAC is a 16 bit converter ($10V * 10000 / 2^{15} \sim 3V$). The output is the encoder measurement. It is the response of the velocity servo loop and it is a ramp in a first approximation, i.e. the integration of the input step.

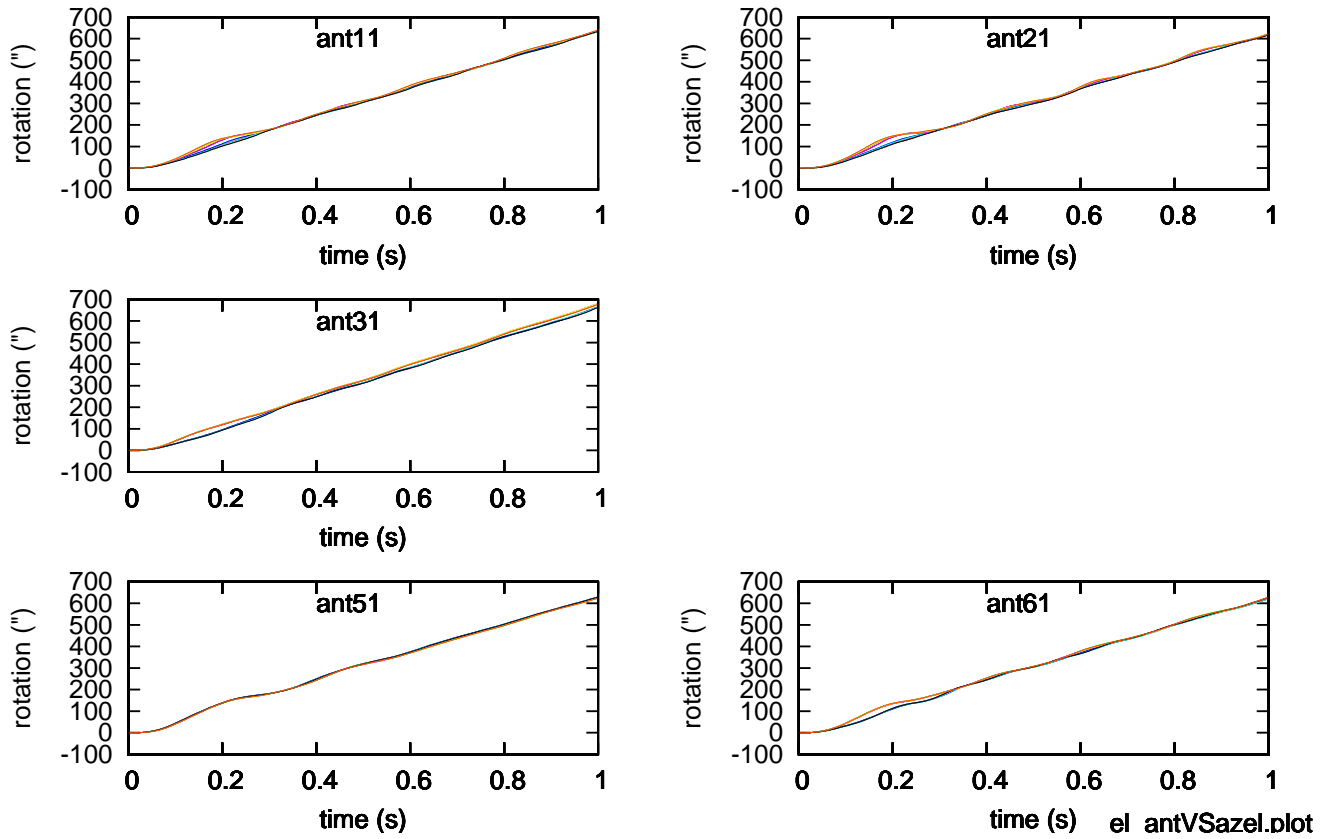
Az(t), step vx= ± 10000 . Az=0,-90deg, El=20, 45, 70deg



The responses are very little dependent on the azimuth and, above all and surprisingly, on the elevation.

1.2 Elevation step response

El(t), step vx= \pm 10000. Az=0,-90deg, El=20, 45, 70deg



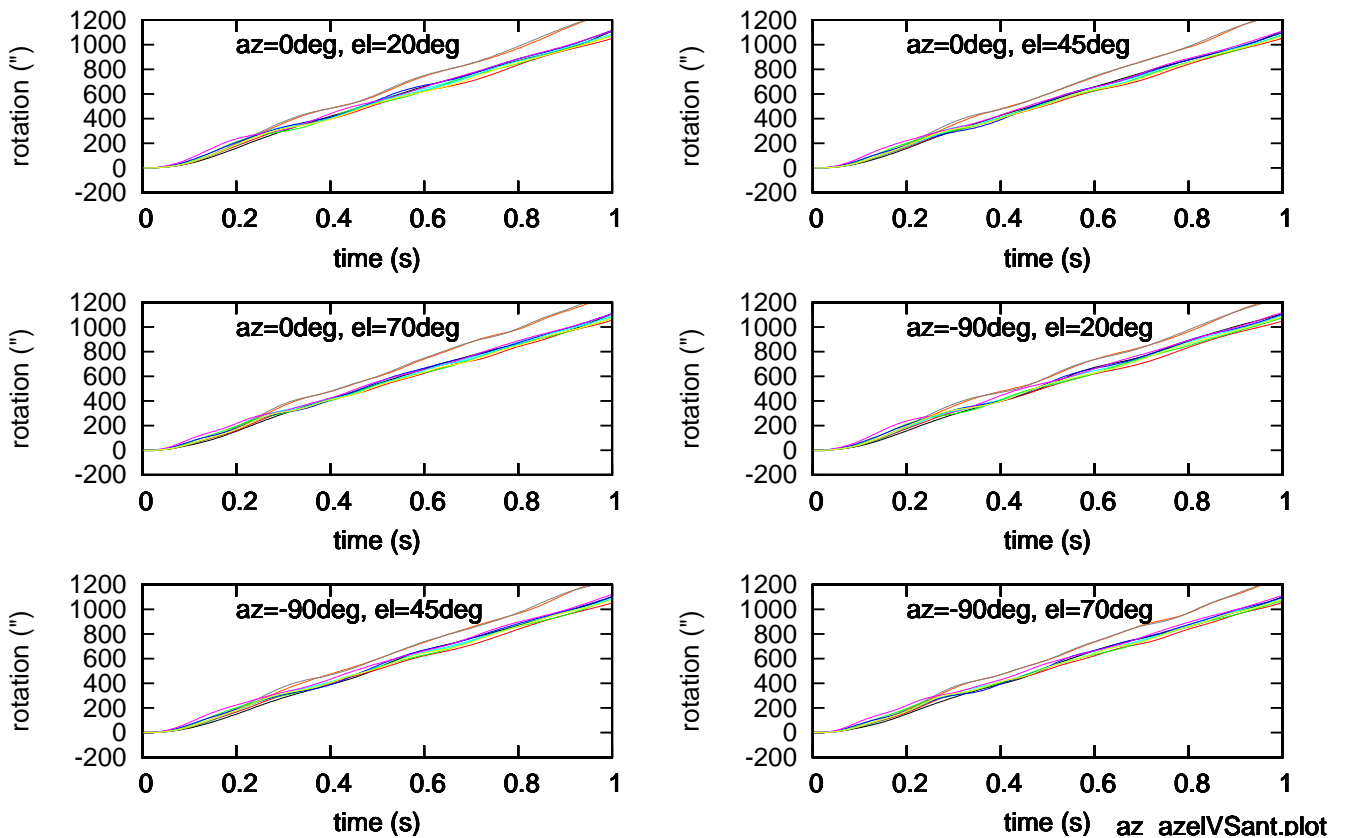
The responses are very little dependent on the azimuth and on the elevation.
 For some antennas one can see differences at the beginning of the ramp, around 0.2s, between positive and negative step requests, i.e. between upward and downward motions. It shows some imbalance of the cabin+dish.

2 Velocity servo loop step response

Versus the antennas.

2.1 Azimuth step response

Az(t), step vx= ± 10000 . ant*1



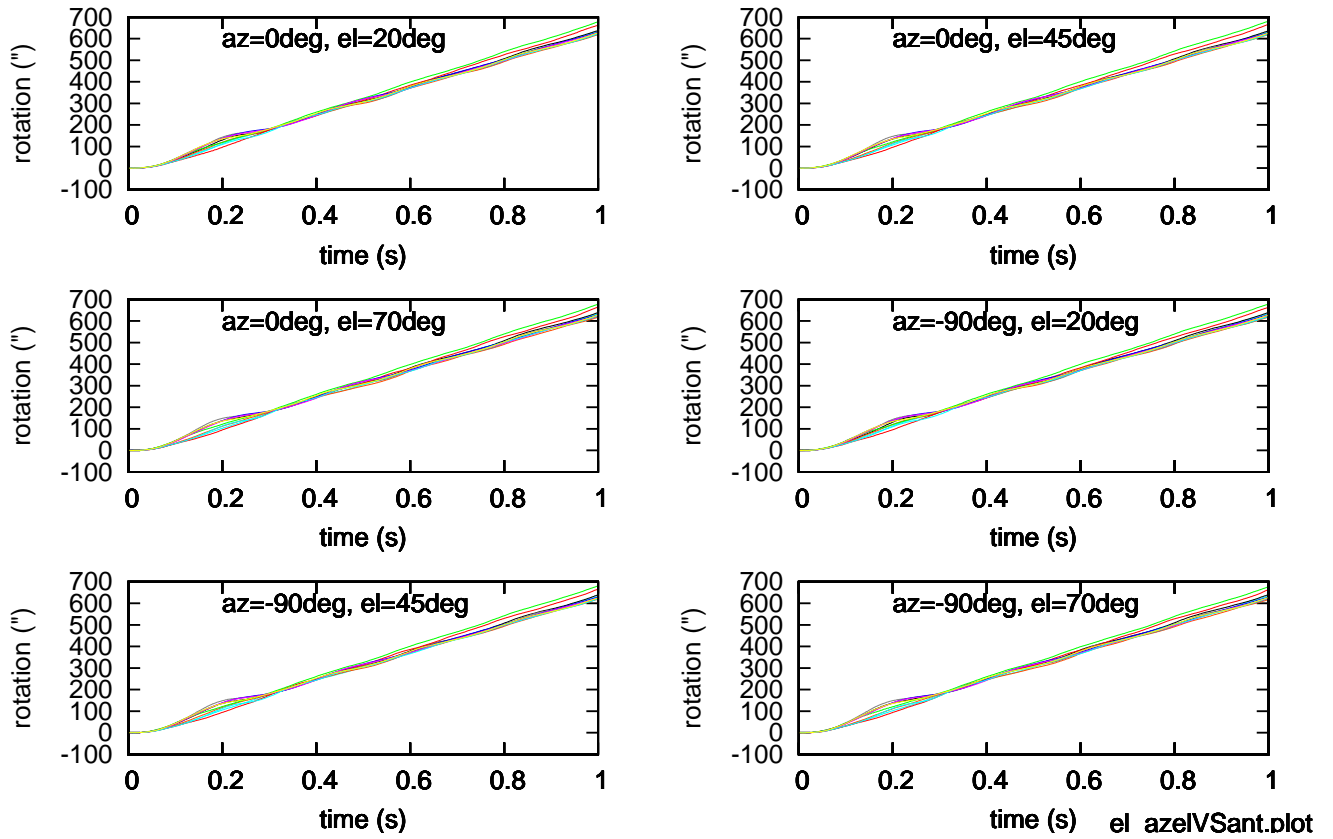
For antennas 1, 3, 5 and 6 the overall gain of the velocity loop is 365 $^{\circ}$ /s/V.

For antenna 2 the gain is 427 $^{\circ}$ /s/V, 17% higher.

Evaluation done with a 16bit DAC, $\pm 2^{15} = \pm 10V$.

2.2 Elevation step response

El(t), step vx= \pm 10000. ant*1



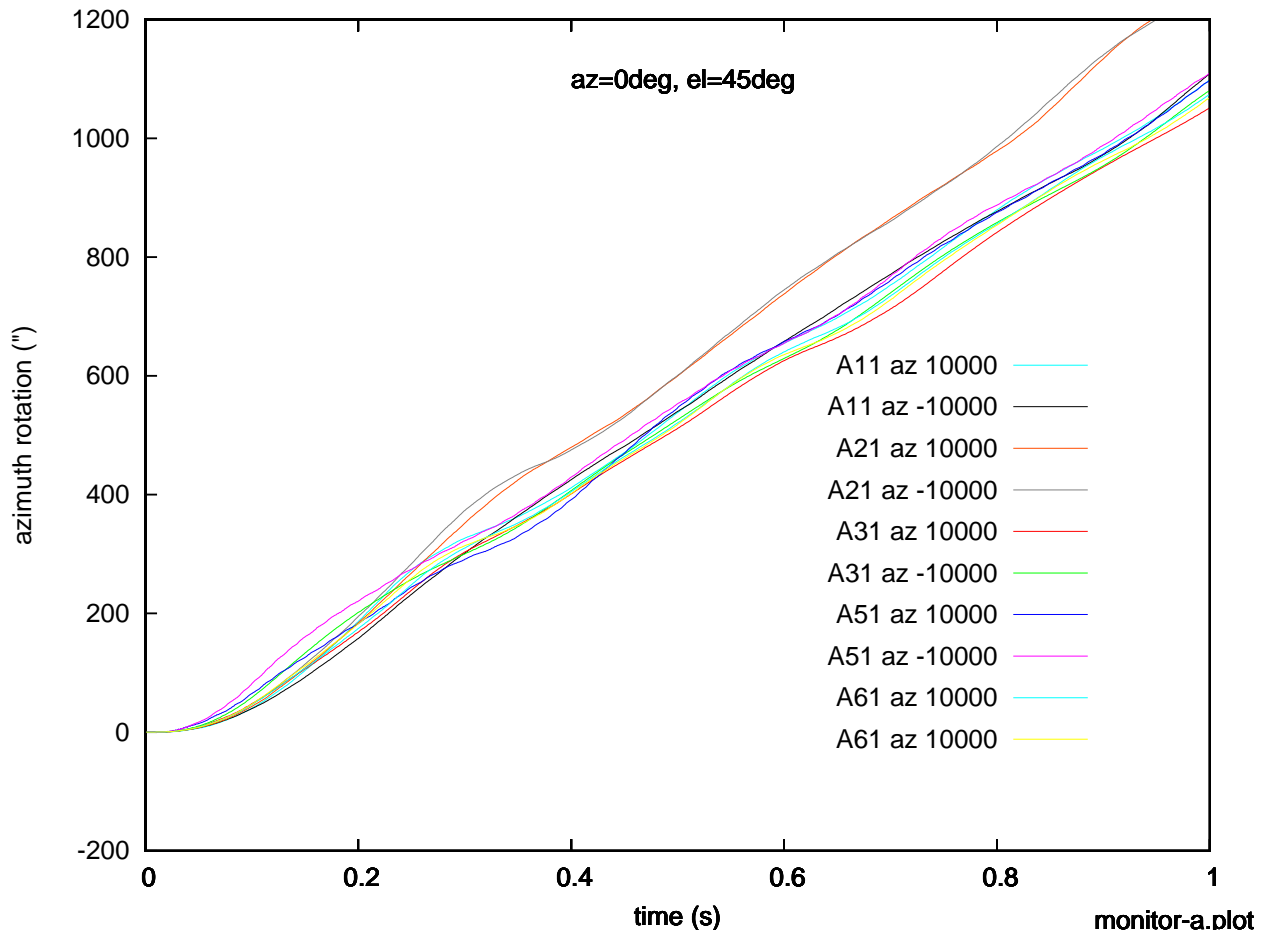
It can be noted some dispersion on the static gains (the slopes of the ramps), in particular around 0.2s and between upward and downward movements.

This difference between positive and negative requested steps implies non-linearity and as consequence, extra phase shift.

3 Frequency analysis

As the ramps are independent of the starting positions in a first approximation, the output curves resulting of positive and negative steps of 3V for az=0deg (south) and el=45deg are analyzed hereafter. To be complete the same study should be performed for different step levels in case, as it is probable, the output is not linear due in particular to the frictions.

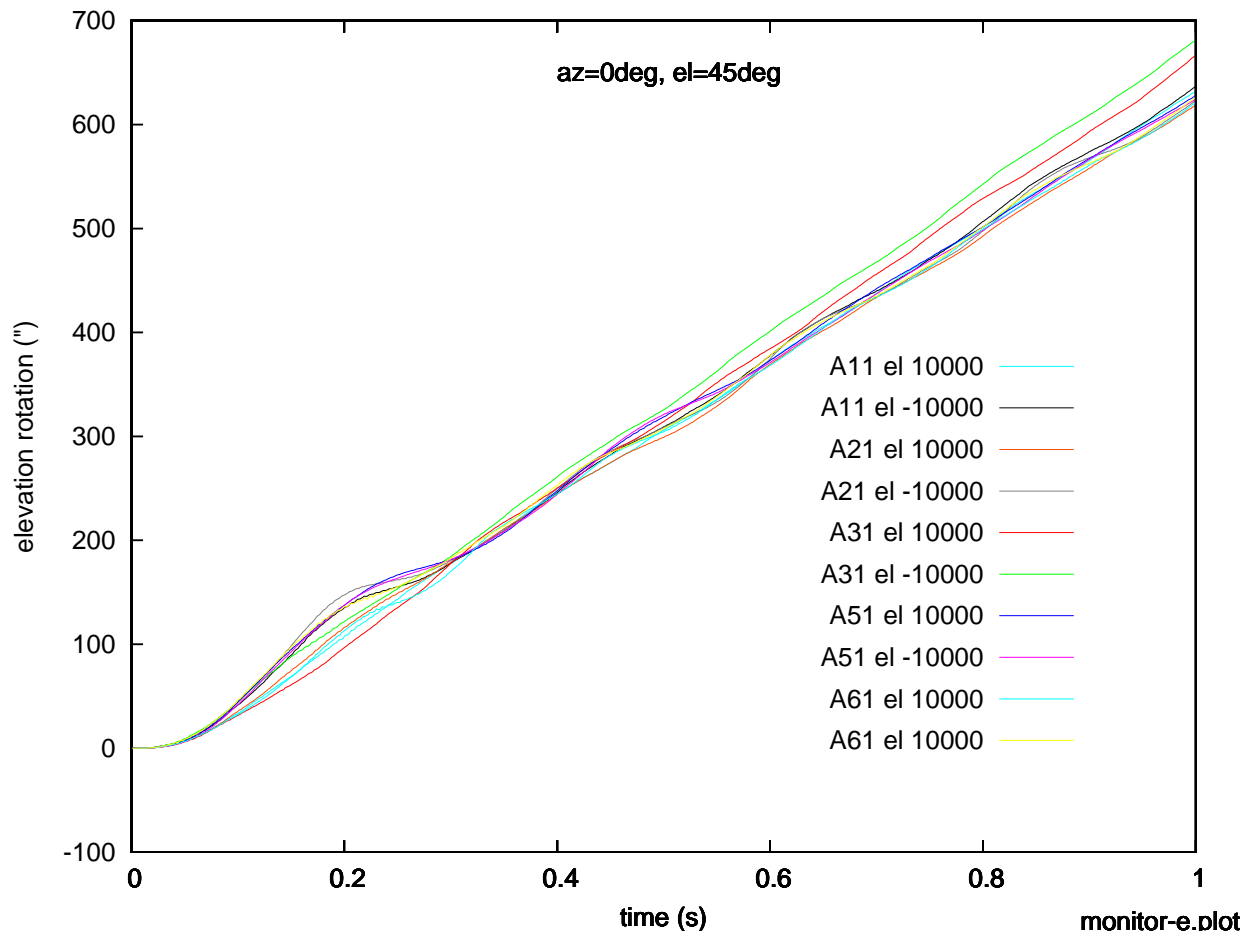
3.1 Azimuth:



There is a position measurement roughly every ms.
 The position resolution is the encoder resolution that is 1/10".
 For each data sample, 2 numbers are recorded: The exact time in microseconds and the position in encoder units.

Here we see clearly the particular case of antenna 2

Elevation:



Here we see the effect of the antenna imbalance in elevation.
 It is clear that a frequency analysis would require getting the same ramps, except the sign, for positive and negative input steps.

3.2 Bode Analysis. Data preprocessing:

Recorded data needs to be re-sampled before a Fast Fourier transform may be applied.

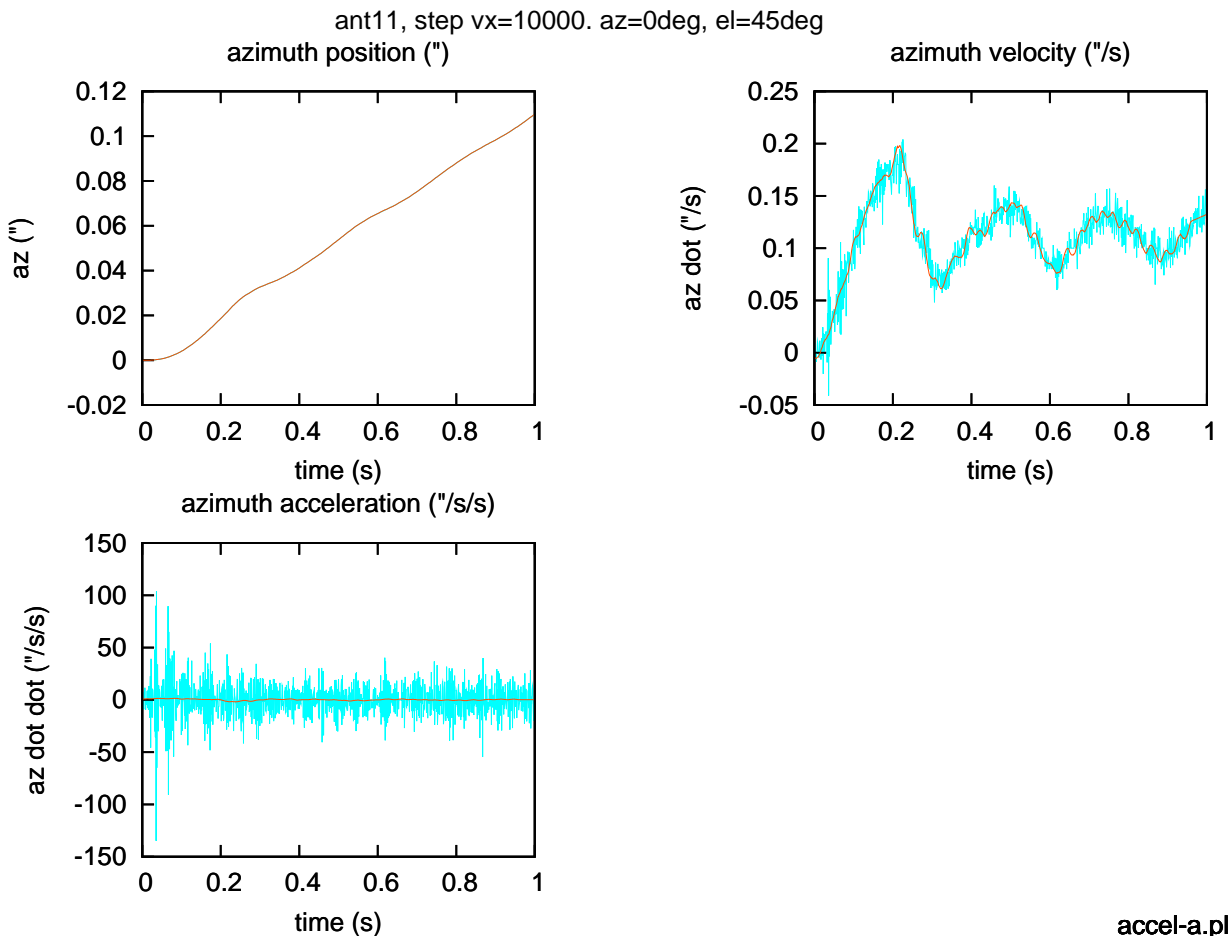
The recording duration of one second is re-sampled into 1024 periods.

Only the acceleration, i.e. the output second derivative, with a theoretical exponential decrease, may be considered as a potential candidate to a FFT.

As it can be seen on the next plots, a derivation process is very noisy.

A sliding polynomial fit on 40 consecutive samples for a 4th degree polynomial is a method for rejecting the high frequencies of the 2nd derivation and for the re-sampling interpolation.

However we will see that the FFT gives practically the same results, for the low frequencies, for an acceleration directly numerically calculated and thus, noisy, or with an acceleration deduced from a filtering process.



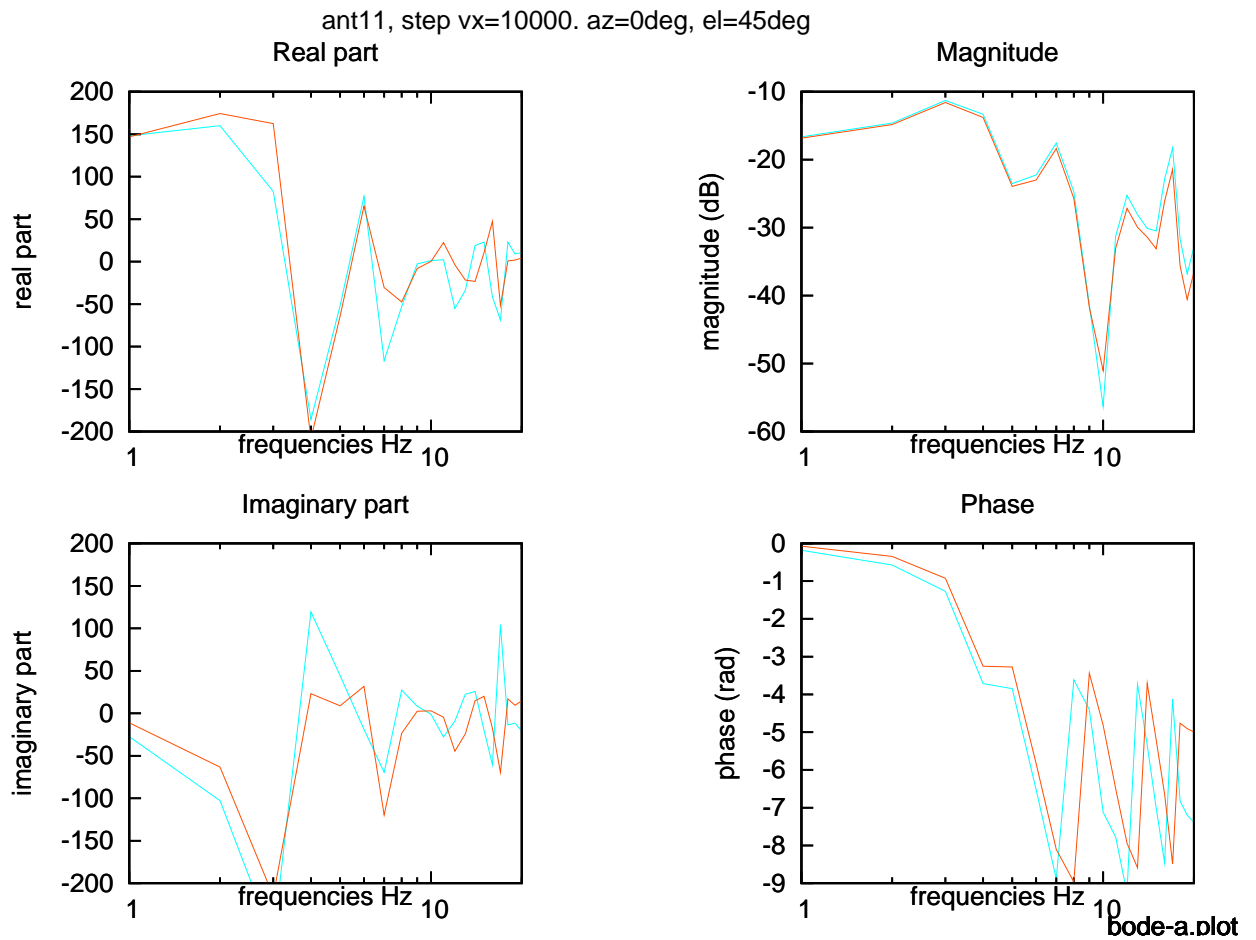
accel-a.plot

The blue lines are the normalized azimuth position and the numerically calculated 1st and 2nd derivatives. The “normalized azimuth position” means the position expressed in arc-second divided by the input step amplitude (+-10000).

The red lines are the filtered and interpolated variables, position, velocity and acceleration.

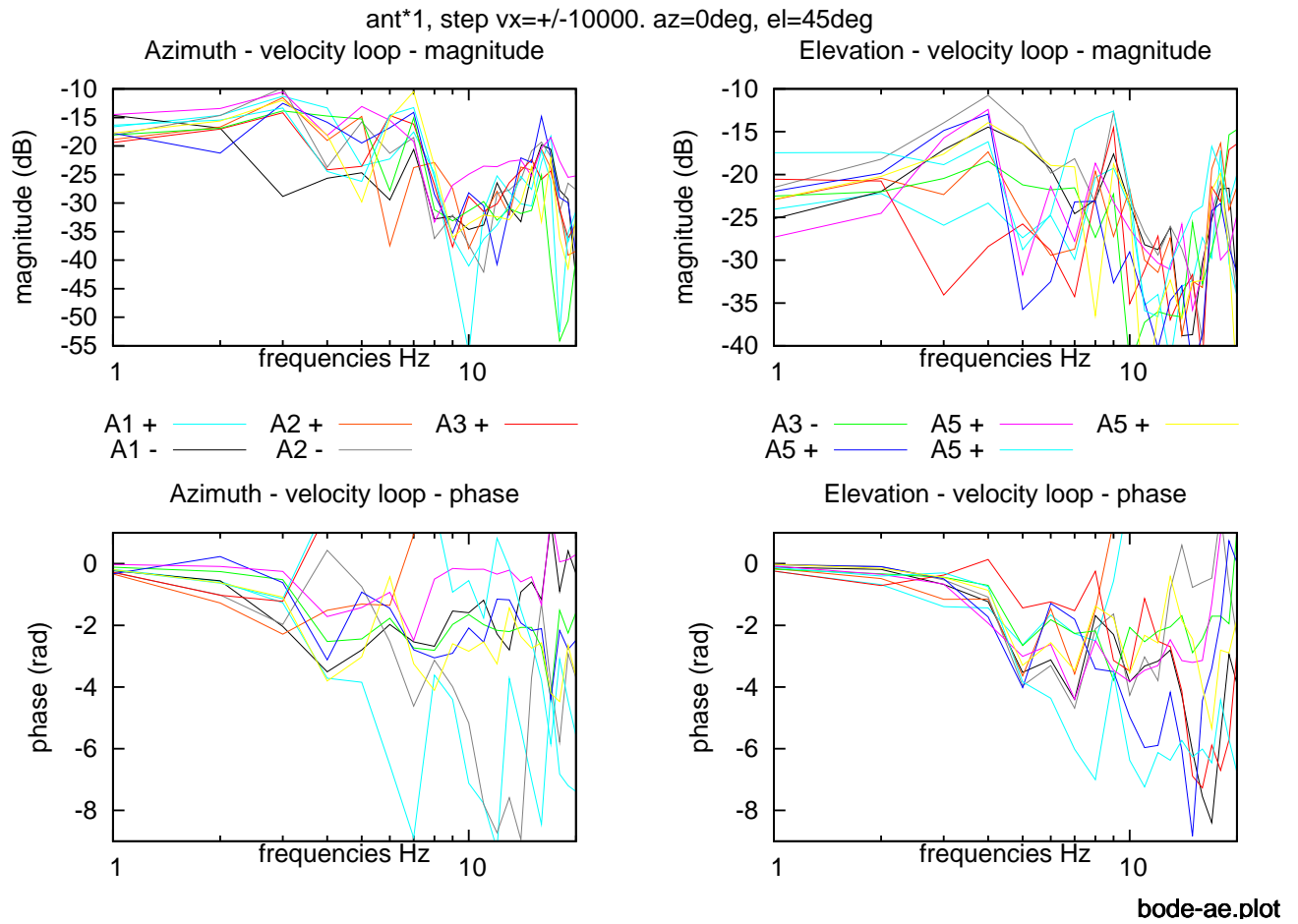
3.3 Transfer function

Comparison of the FFT of the 2 calculated accelerations for the very low frequencies (up to 20Hz)



The 2 FFTs are very similar. The noise resulting of the numerical acceleration calculation does not convey any information at low frequency.
 Here a resonance around 7Hz is visible (peak on the magnitude and phase shift of more than PI).

The same processing is applied to all antennas for the azimuth and the elevation.
 The fact that a difference is visible between the positive step and the negative one, shows the non-linearity of the transfer function and the limits of the Bode analysis.



The interpretation is limited by the lack of data at low frequency. A longer recording would be an advantage.
 However a resonance is visible around 7Hz in azimuth and 9Hz in elevation.

4 Ideal case analysis

The velocity loop transfer function (velocity versus DAC input command) should be of the form

$$\dot{Y} = \frac{G}{1 + \frac{s}{\omega}} X$$

with \dot{Y} the axis velocity and X the DAC input

G Is the static gain in "/s/bit (In principle it could be in rad/s/V, knowing that $X=2^{15} == 10V$)
 ω is the corner or cutoff frequency in rad/s

The input is a step function:

$$X = U / s \quad U \text{ is a constant (here 10000)}$$

and the position output looks like:

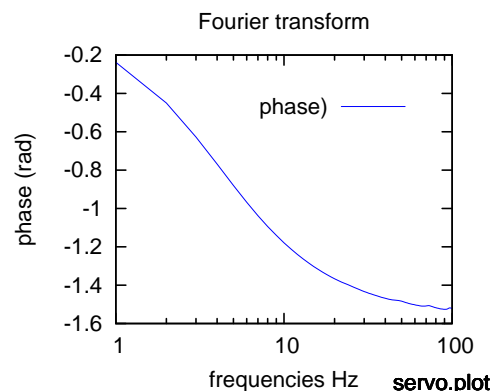
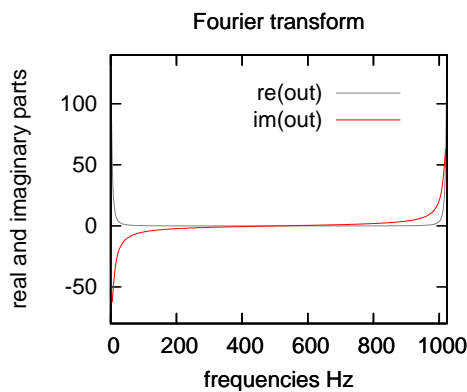
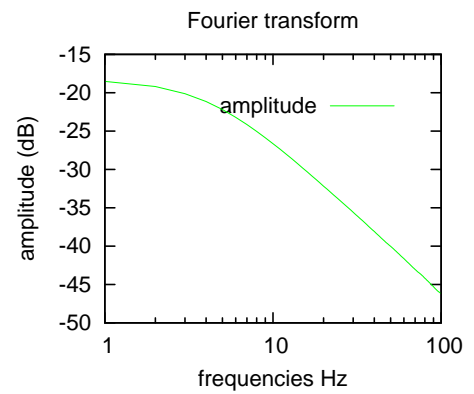
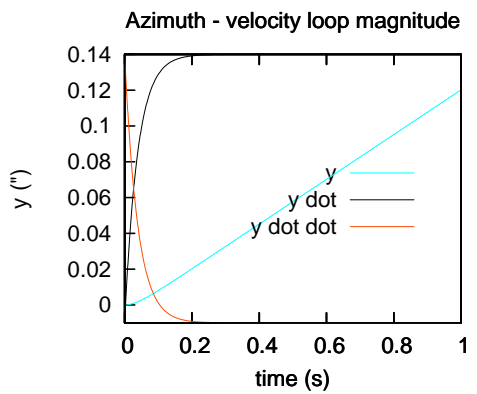
$$Y = \frac{G}{\omega} \frac{U}{1 + \frac{s}{\omega}}$$

In the time domain

$$\frac{\ddot{y}(t)}{U} = G \omega e^{-\omega t}$$

$$\frac{\dot{y}(t)}{U} = G - G e^{-\omega t}$$

$$\frac{y(t)}{U} = -\frac{G}{\omega} + Gt + \frac{G}{\omega} e^{-\omega t}$$



with $G = .1250 \text{ "/s/bit} == 409.6 \text{ "/s/V}$
 and $\omega = 26 \text{ rad / s} == 4.13 \text{ Hz}$

servo.plot

5 Position servo loops

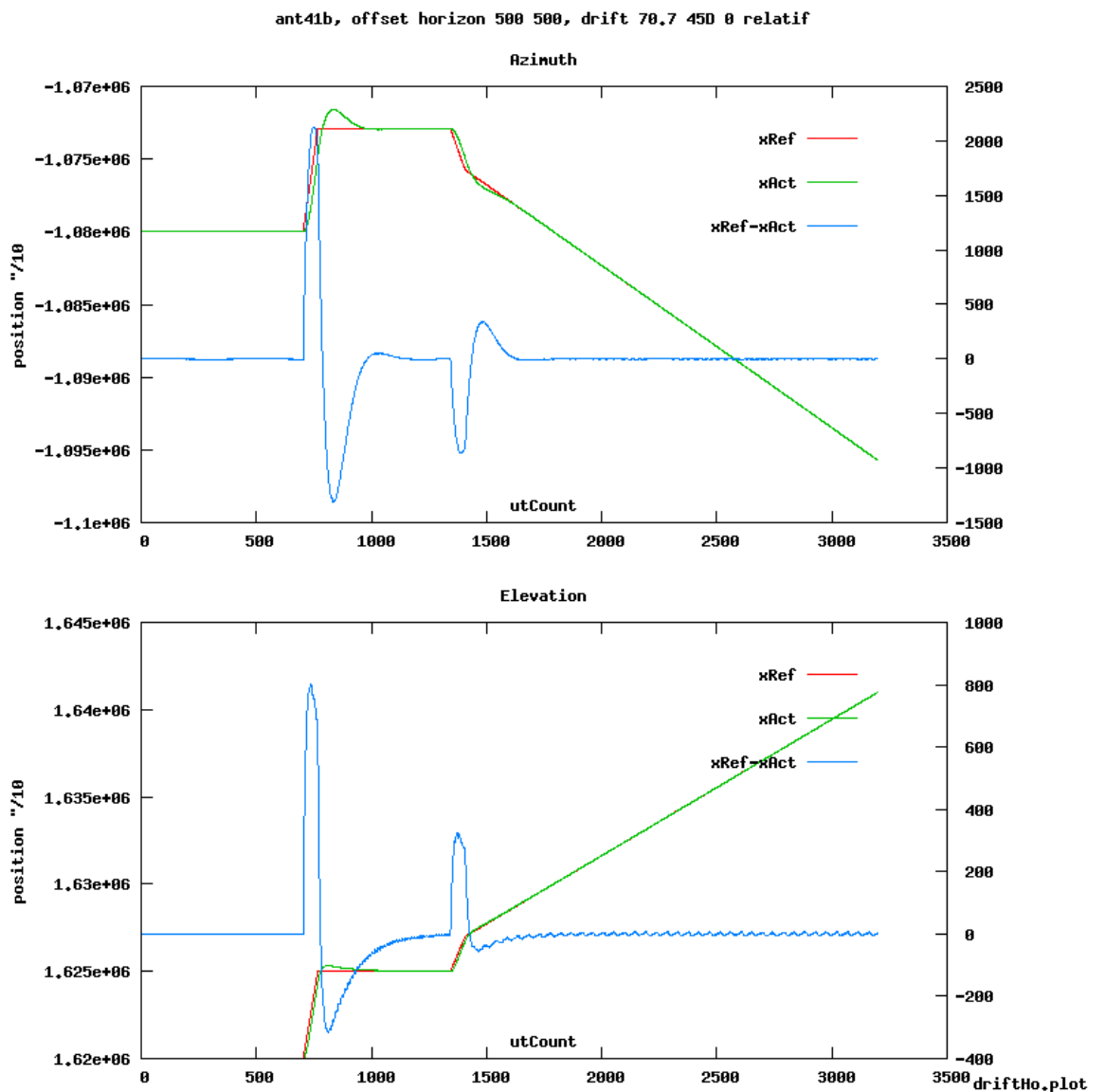
Both axis are studied from tracking on a fixed position in the horizon coordinate system, then an offset compatible with the position tracking is applied. Finally a drift is added on top. The commands from bure1b looks like this for antenna x, from 1 to 6:

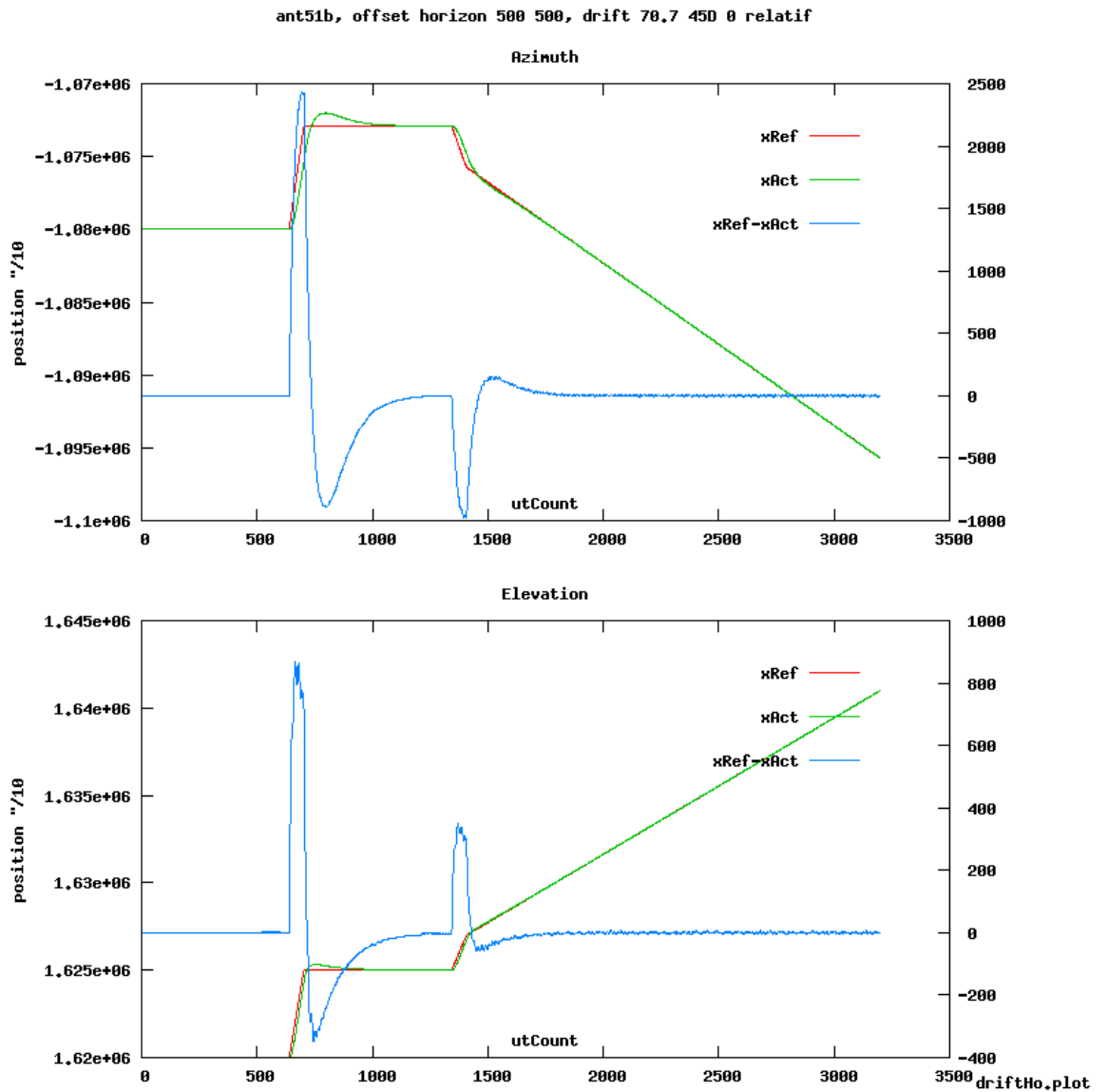
```

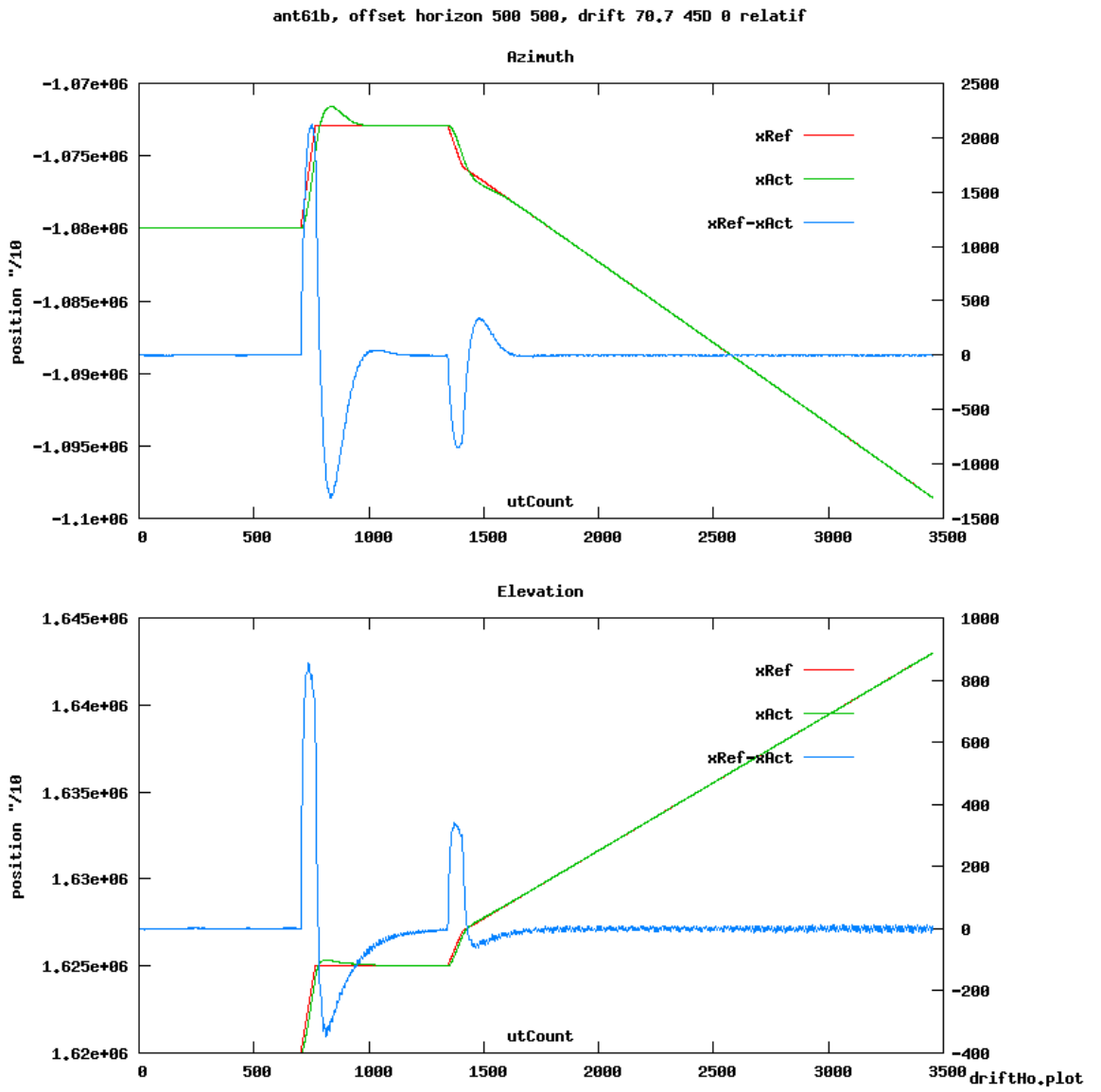
coo antenna x horizon -30D 45D
off antenna x horizon 500 500
dri antenna x horizon 70.7 45D 0 relatif
    
```

There is a delay of 3s between the moment the command dri is issued and the moment the command is applied. It is explained by the periodic execution of interp which send the command to the antenna, and all the coordinate computation calculated some time ahead in the pedestal processor.

Due to some imperfections in the scripts for the tests, the measurements do not yet cover antennas 1, 2 and 3 in this horizon tracking mode.

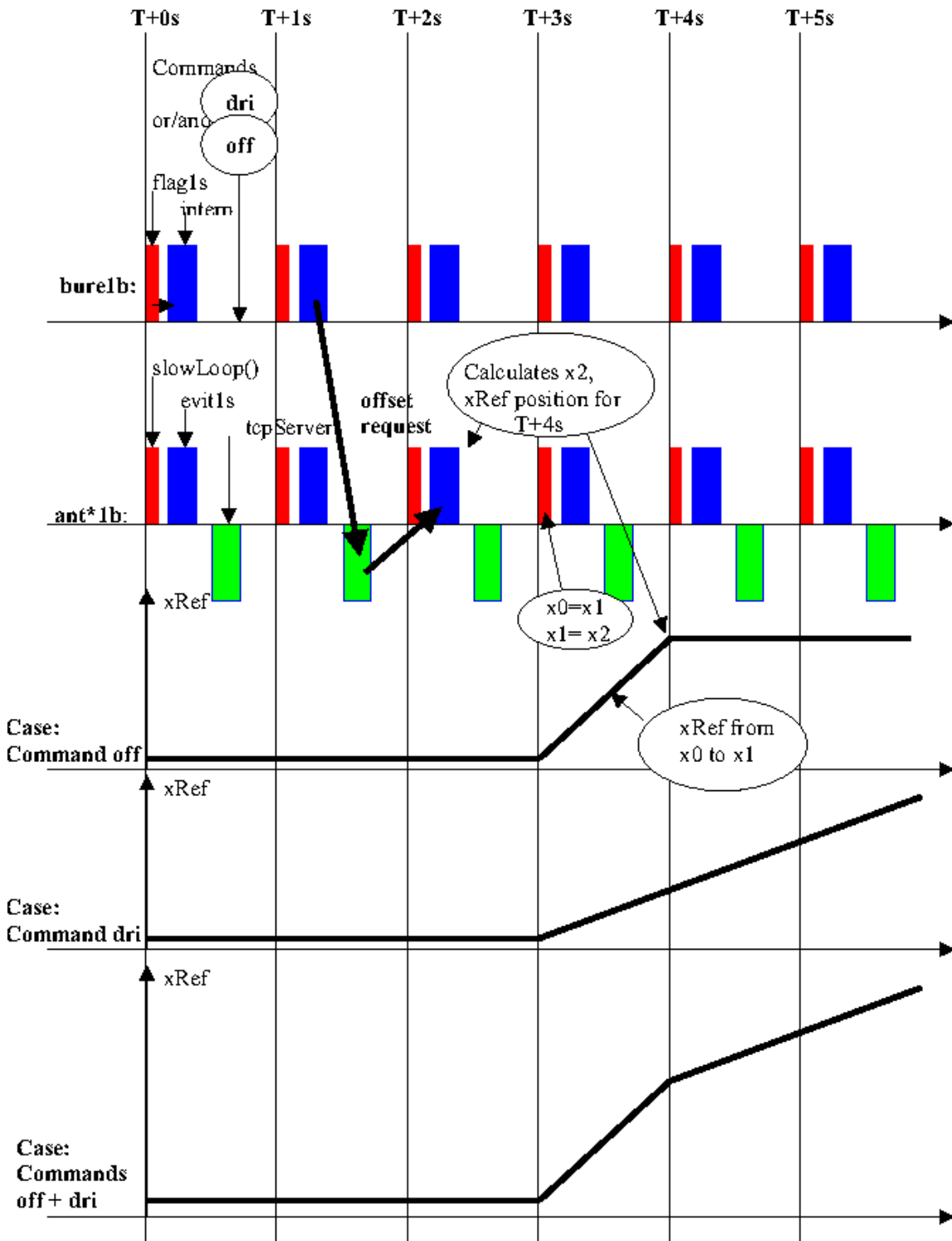


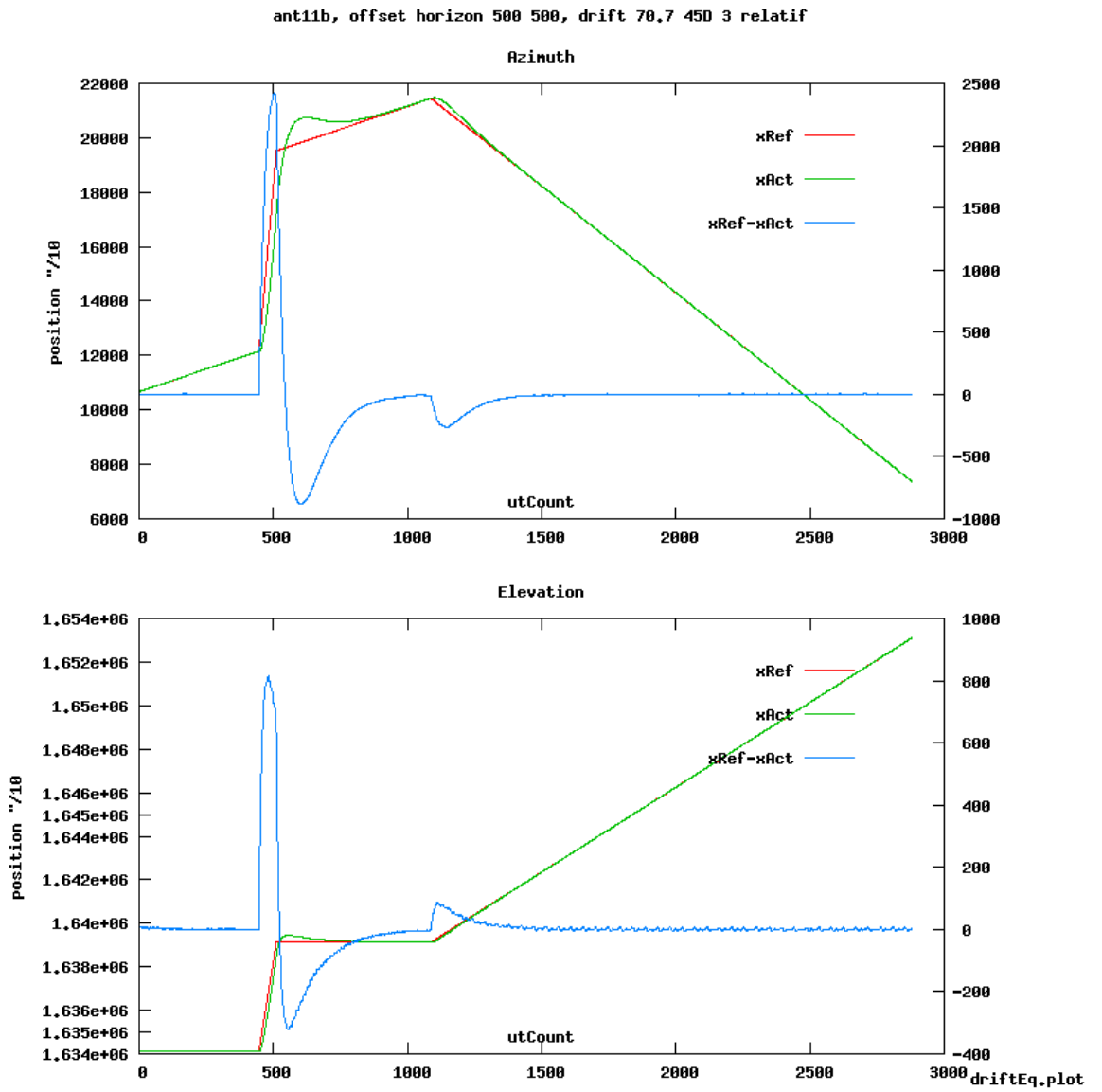


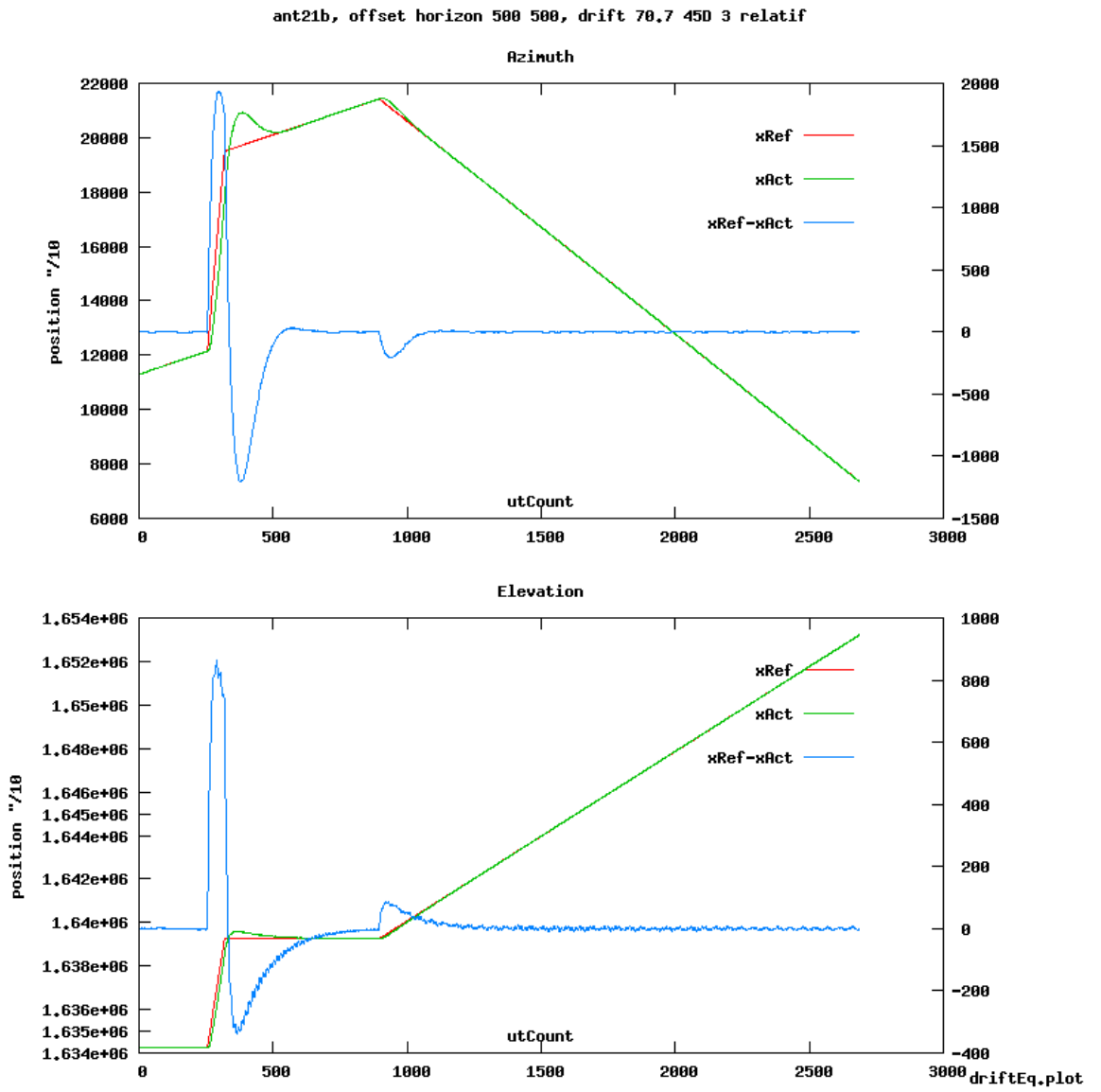


Both axis are studied while tracking on a fixed position in the equatorial coordinate system, then an offset compatible with the position tracking is applied. Finally a drift is added on top. The commands from bure1b looks like this for antenna x, from 1 to 6:

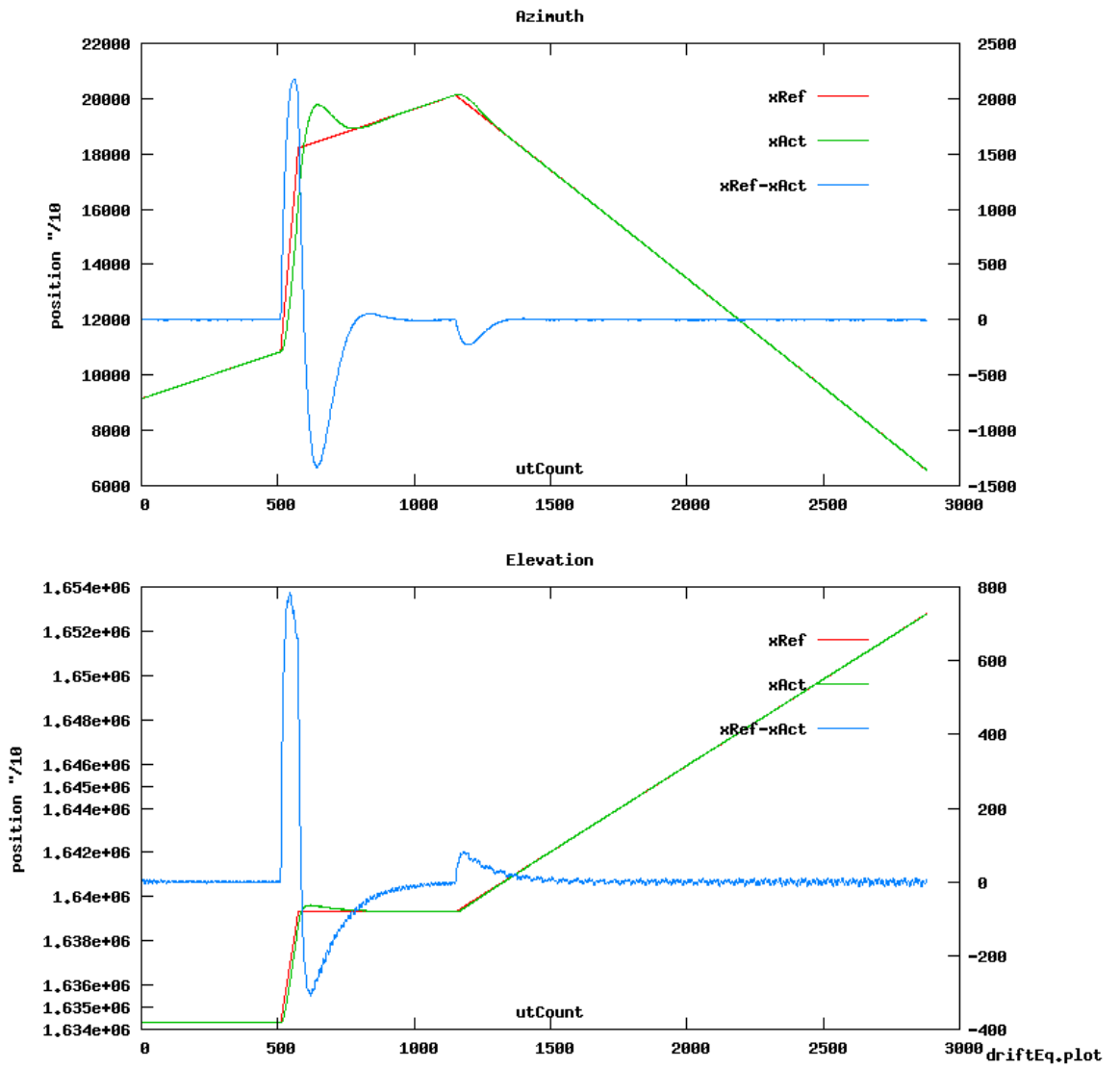
```
coo antenna x equatorial <sideral time> 0
off antenna x horizon 500 500
dri antenna x horizon 70.7 45D 3 relatif
```

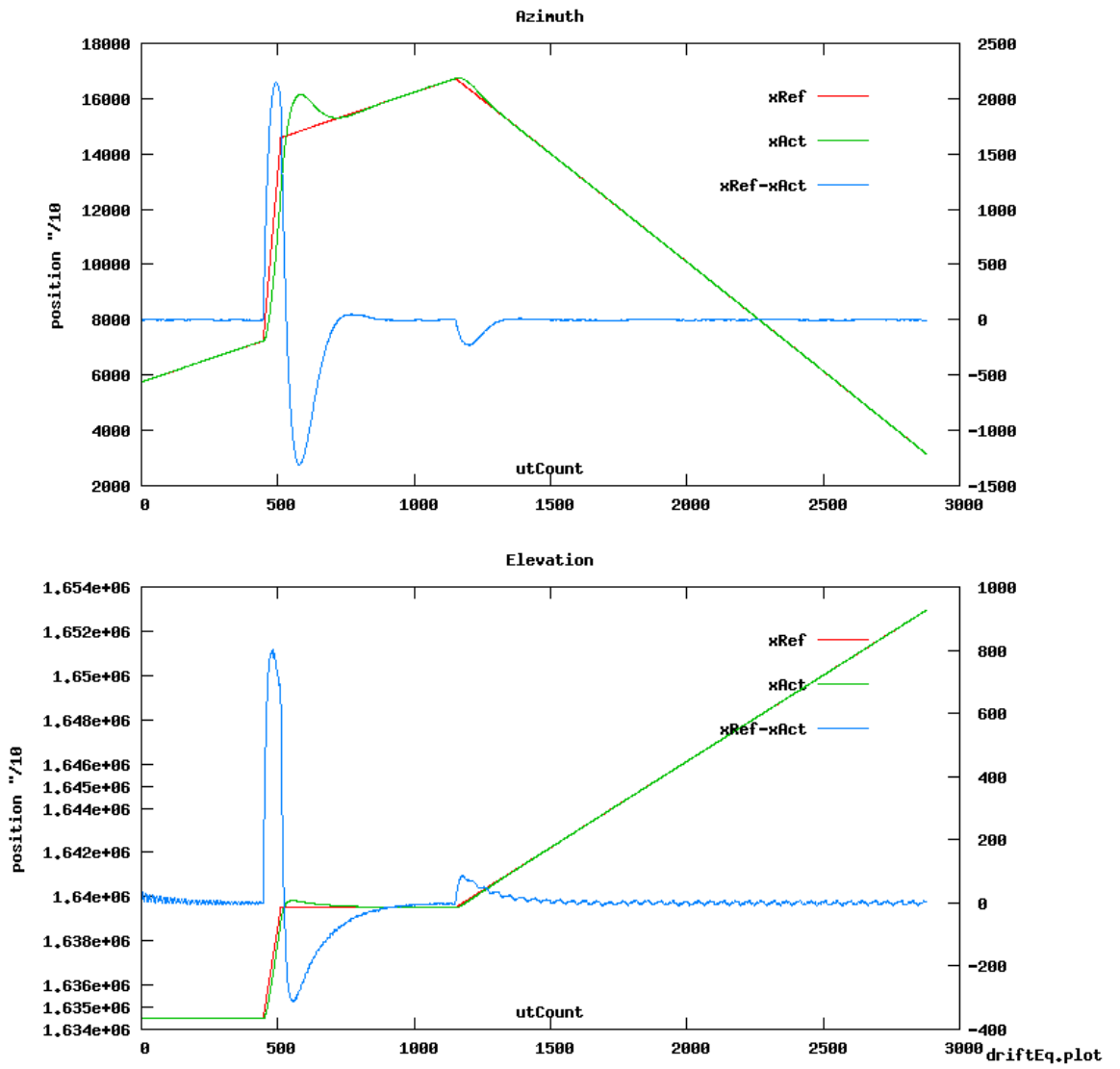


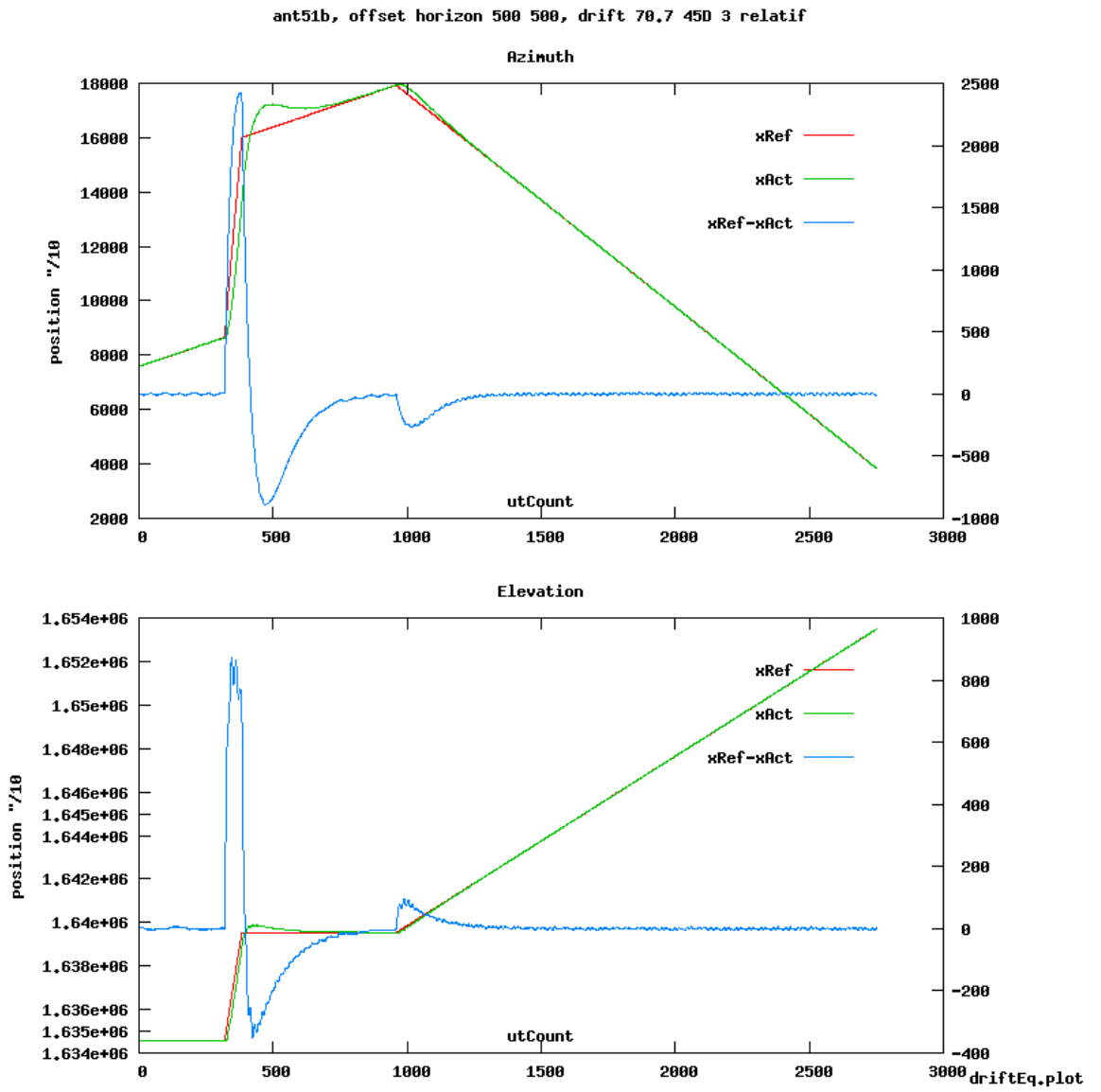


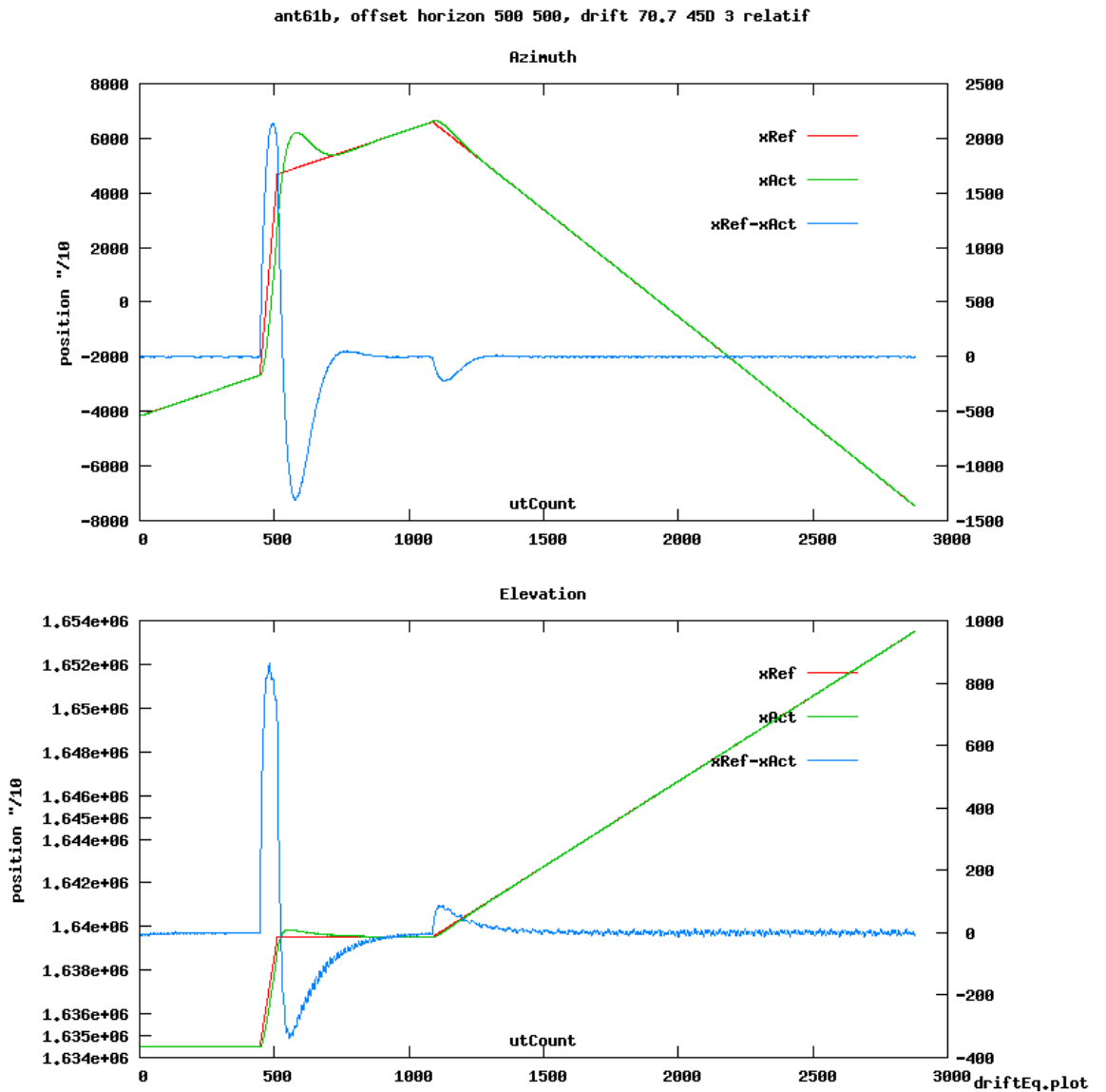
ant31b, offset horizon 500 500, drift 70.7 450 3 relatif



ant41b, offset horizon 500 500, drift 70.7 450 3 relatif

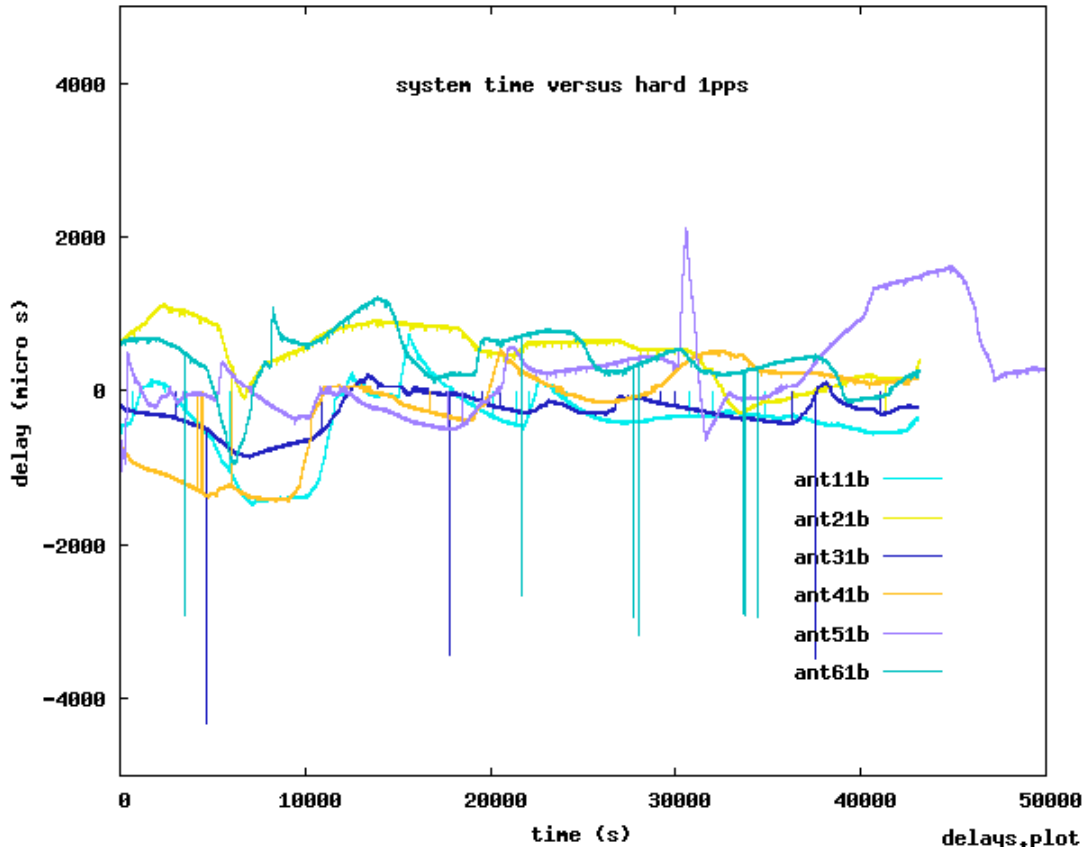






6 System time delays

The next plot shows, for all antennas, the variation of the system time at the 1pps time bus event time, from just after a reboot of the pedestal processor and for around 12 hours. In other words at each interrupt generated by the time bus signal, i.e. every second, the fraction of the second of the system time synchronized on the ntp system is recorded. The fraction of the second is expressed in micro seconds. We see that the time difference does not vary more than ± 2 ms, from the beginning, which is small enough to guarantee the identification of each period of the fast (64Hz) loop by just reading the system time.



Reminder:

On each antenna the command `ntpq -p` shows the current status of the ntp synchronization.

```
ant11b root:~ # ntpq -p
  remote      refid  st t when poll reach  delay  offset  jitter
=====
*clockb.iram.fr .TRUE.  1 u  41  256  377   0.339   0.452   0.339
```

The current estimated delay, offset and dispersion of the peer (clockb) are expressed in milliseconds.

The server ntpd is started at boot time with the command:

```
/usr/local/bin/ntpd -A -c /etc/ntp-broadcast.conf &
```

- A Disable authentication mode
- c Specify the name and path of the configuration file

```
ant11b root:/home/introot # cat /etc/ntp-broadcast.conf
broadcastclient           # broadcast client
driftfile /tmp/ntp.drift
```